

The Optimal Steady-State Control Problem

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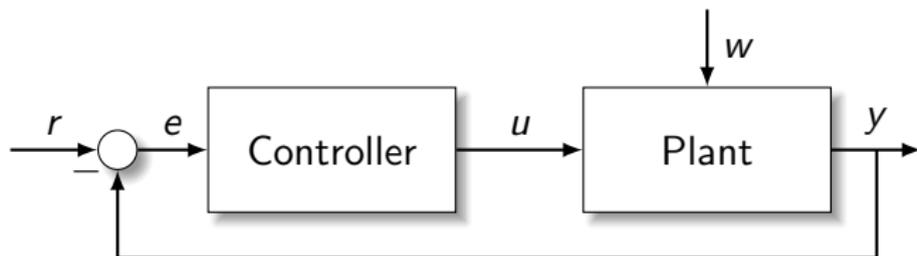
Liam S. P. Lawrence
University of Waterloo



Enrique Mallada
John's Hopkins Univ.

Control Systems 101

- Prototypical feedback control problem is **tracking** and **disturbance rejection** in the presence of non-negligible **model uncertainty**

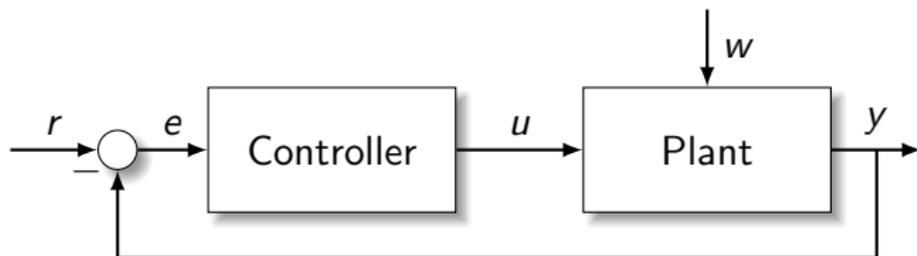


- **Exact robust asymptotic tracking** achieved if loop gain *“incorporates . . . a suitably reduplicated model of the dynamic structure of the exogenous signal”*

How is the reference r being determined?

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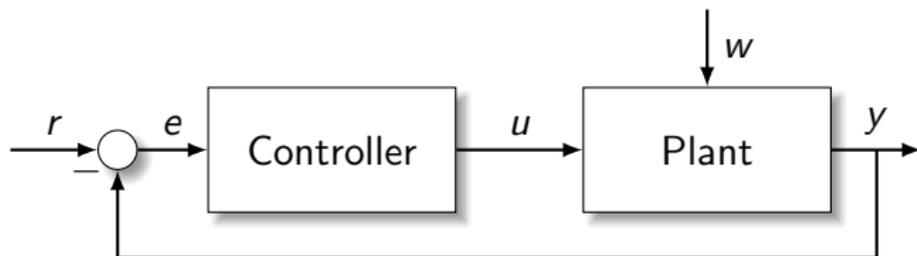


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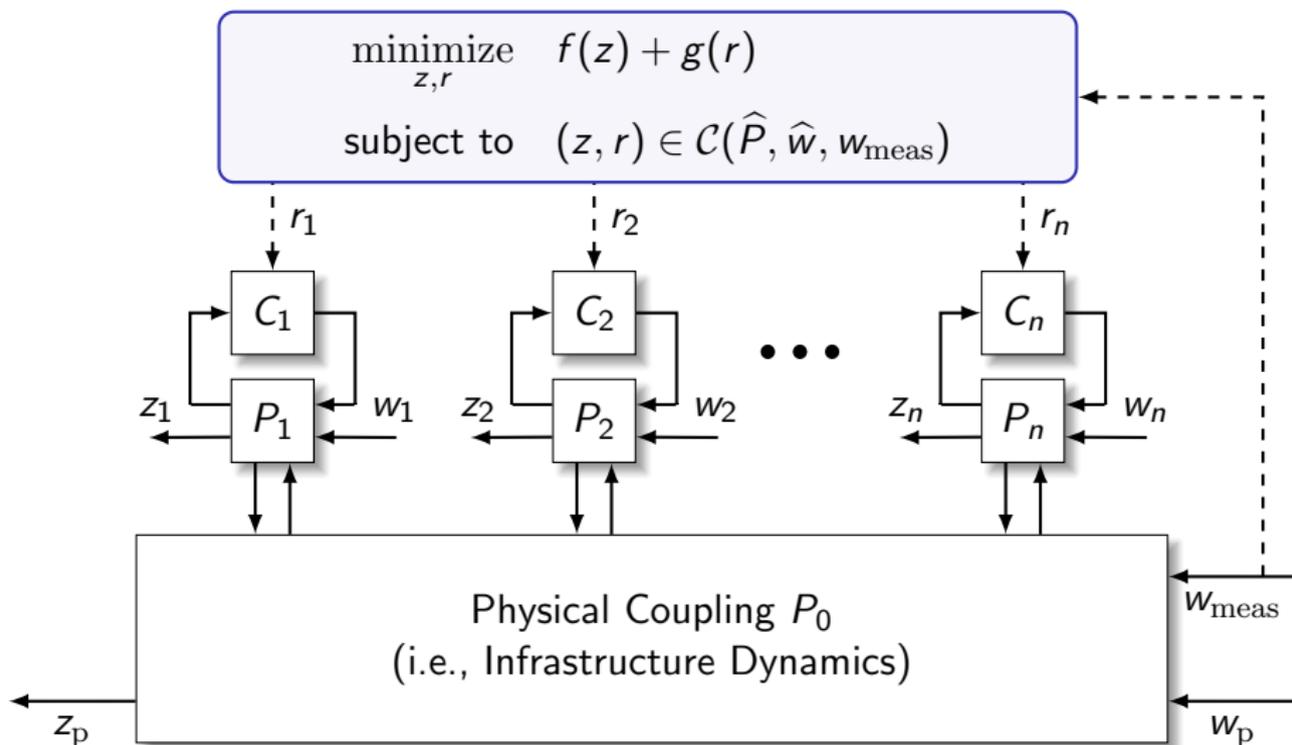
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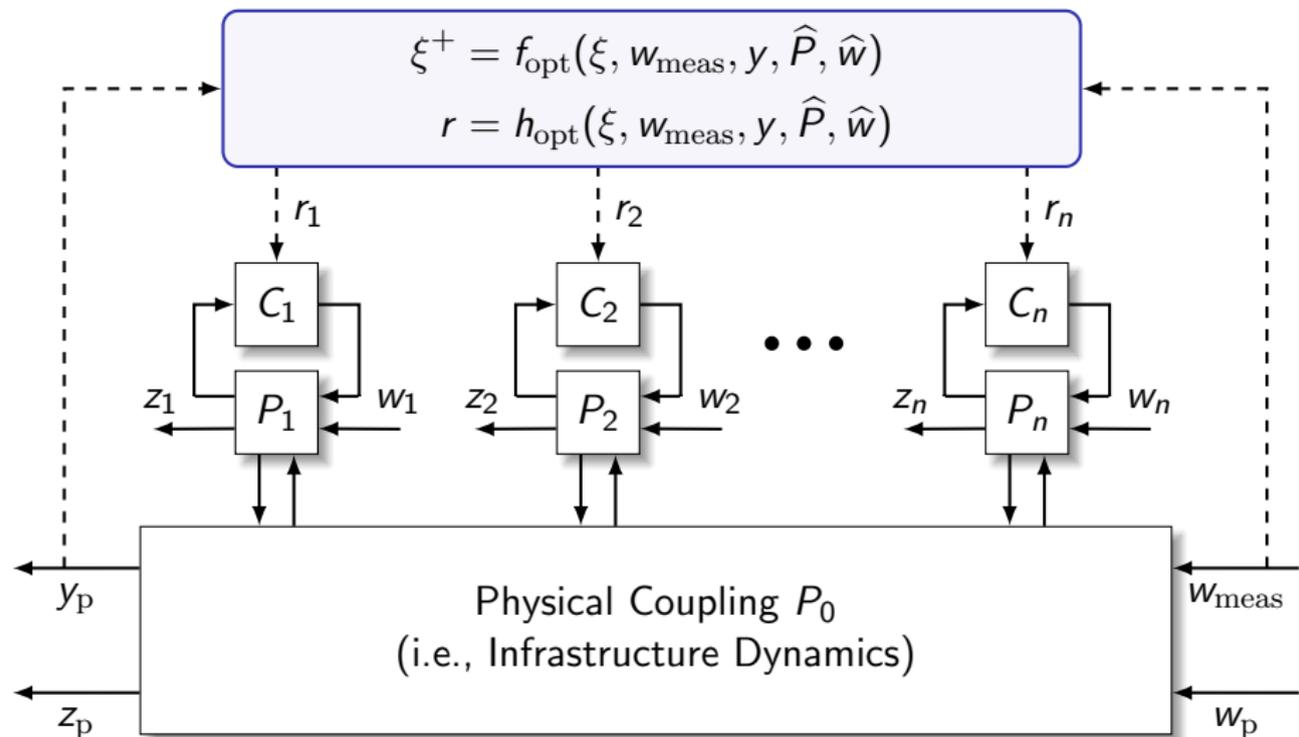
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Feedforward Optimization of Large-Scale Systems



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Feedforward vs. Feedback Optimization

Property	Feedforward	Feedback
Setpoint Quality	≈ Optimal	≈ Optimal
High-Fidelity Model	Crucial	Not crucial
Robustness	No	Yes
Feedback Design/Analysis	Unchanged	More difficult
Computational Effort	Moderate	???

MPC: high computational effort, difficult analysis ⇒ Alternatives?

Compared to MPC, if we **give a bit** on trajectory optimality, can we can **gain a lot** on ease of feedback design/analysis and computational effort?

Here is a first cut of such an approach.

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Optimal Steady-State Control Problem Statement

Given:

- 1 a dynamic system model with
 - a *class* of external disturbances $w(t)$
 - an uncertainty specification (e.g., parametric)
- 2 a vector of outputs $y \in \mathbb{R}^p$ of system to be optimized
- 3 an optimization problem in y

Design, if possible, a feedback controller such that

- 1 closed-loop is (robustly) well-posed and internally stable
- 2 the regulated output tracks its optimal value

$$\lim_{t \rightarrow \infty} y(t) - y^*(t) = 0, \quad \forall w, \forall \text{ uncertainties}$$

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LTI-Convex OSS Control: Setup Overview

1 Uncertain LTI dynamics

$$\begin{aligned}\dot{x} &= A(\delta)x + B(\delta)u + B_w(\delta)w \\ y_m &= C_m(\delta)x + D_m(\delta) + Q_m(\delta)w \\ y &= C(\delta)x + D(\delta)u + Q(\delta)w\end{aligned}$$

- δ = parametric **uncertainty**, w = const. **disturbances**
- y_m = system measurements available for **feedback**
- y = arbitrary system states/inputs to be **robustly optimized**

2 a steady-state **convex optimization problem**

$$y^*(w, \delta) = \operatorname{argmin}_{y \in \mathbb{R}^p} \{ f(y, w) : y \in \mathcal{C}(w, \delta) \}$$

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LTI-Convex OSS Control: Setup I

Forced equilibria $(\bar{x}, \bar{u}, \bar{y})$ satisfy

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This defines an **affine set** of *achievable* steady-state outputs

$$\bar{Y}(w, \delta) = (\text{offset vector}) + V(\delta)$$

Note: Due to

- 1 selection of variables $y \in \mathbb{R}^p$ to be optimized, and/or
- 2 structure of model matrices (A, B, C, D, B_w, Q)

it may be that $\bar{Y}(w, \delta) \subset \mathbb{R}^p$

constraint $\iff \bar{y} \in \bar{Y}(w, \delta)$ cannot be ignored

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LTI-Convex OSS Control: Setup II

Desired regulated output $y^*(w, \delta)$ solution to

minimize	$f(y, w)$	(convex cost)
	$_{y \in \mathbb{R}^p}$	
subject to	$y \in \bar{Y}(w, \delta) = y(w, \delta) + V(\delta)$	(equilibrium)
	$Hy = Lw$	(engineering equality)
	$k_i(y, w) \leq 0$	(engineering inequality)

Equilibrium constraints ensure **compatibility** between the plant and the optimization problem

\implies guarantees a steady-state exists s.t. $y = y^*(w, \delta)$.

We want to **track** optimal output $y^*(w, \delta)$

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Optimality Models for OSS Control

An **optimality model** filters the available measurements to robustly produce a proxy error ϵ for the true tracking error $e = y^*(w, \delta) - y$

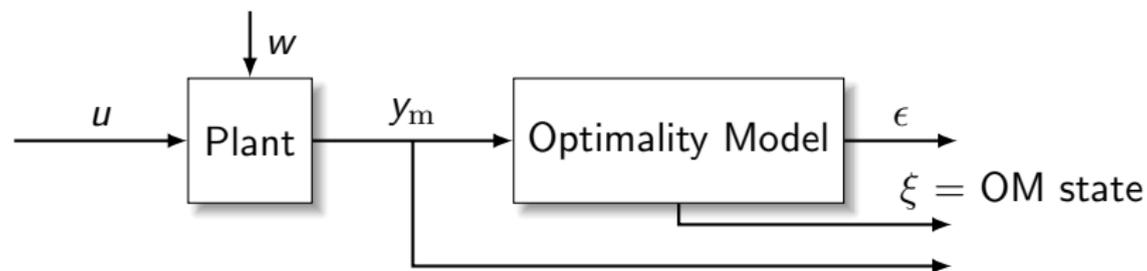


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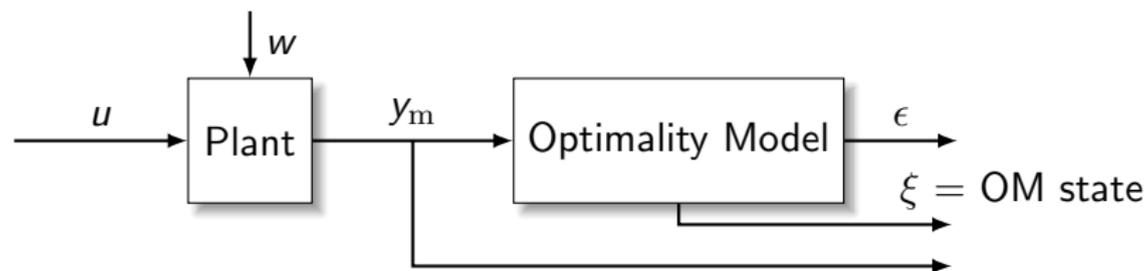


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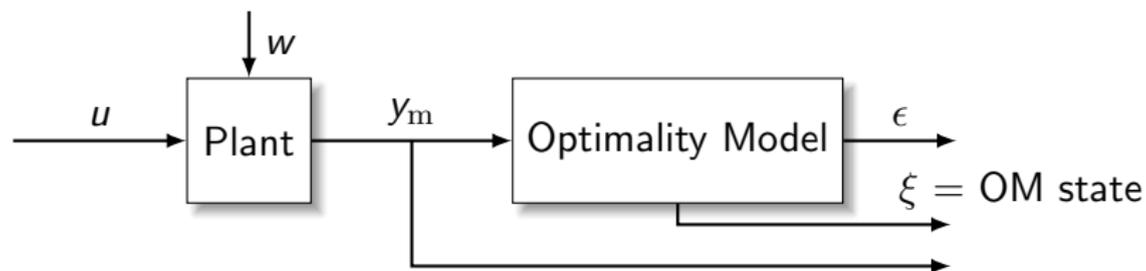


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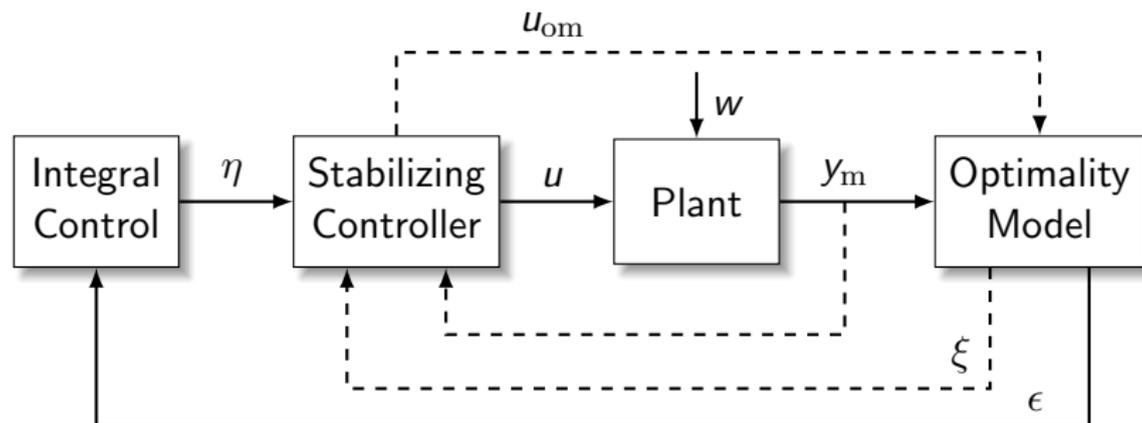


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General Architecture for OSS Control

Optimality model reduces OSS control to output regulation



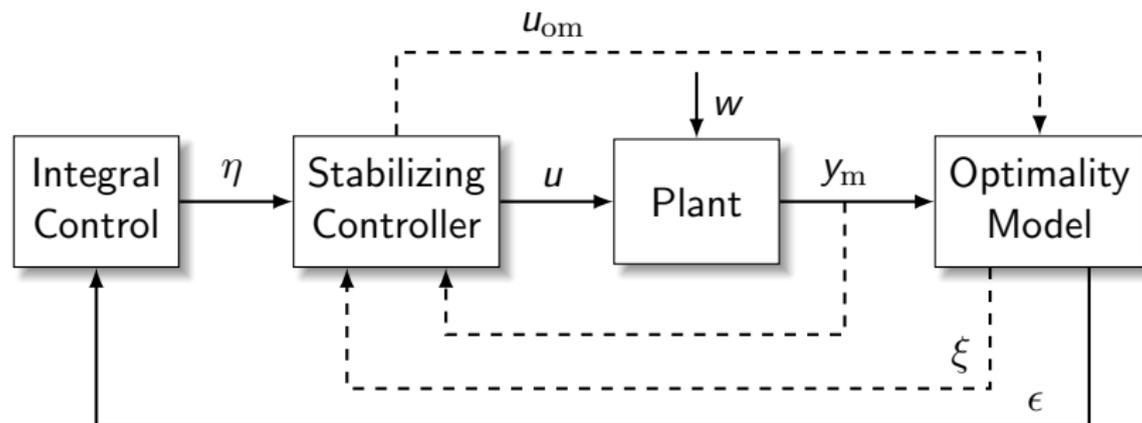
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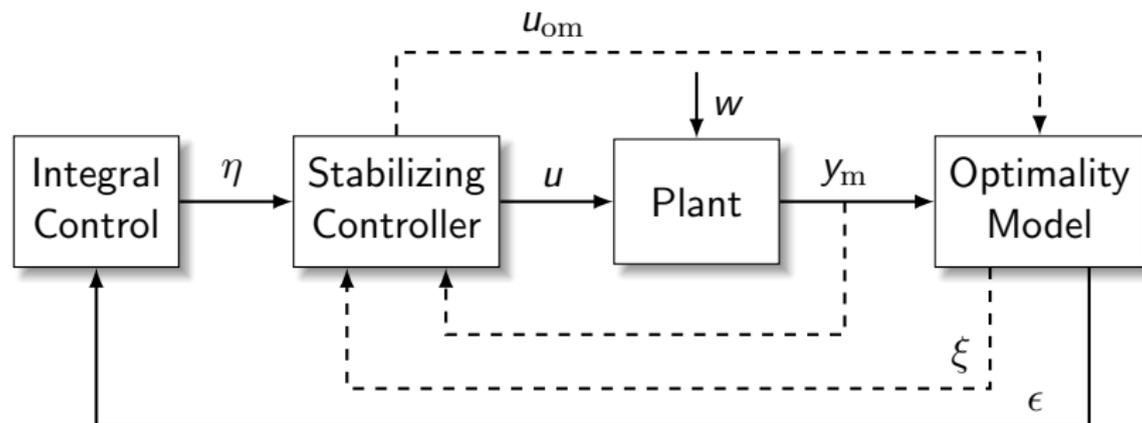
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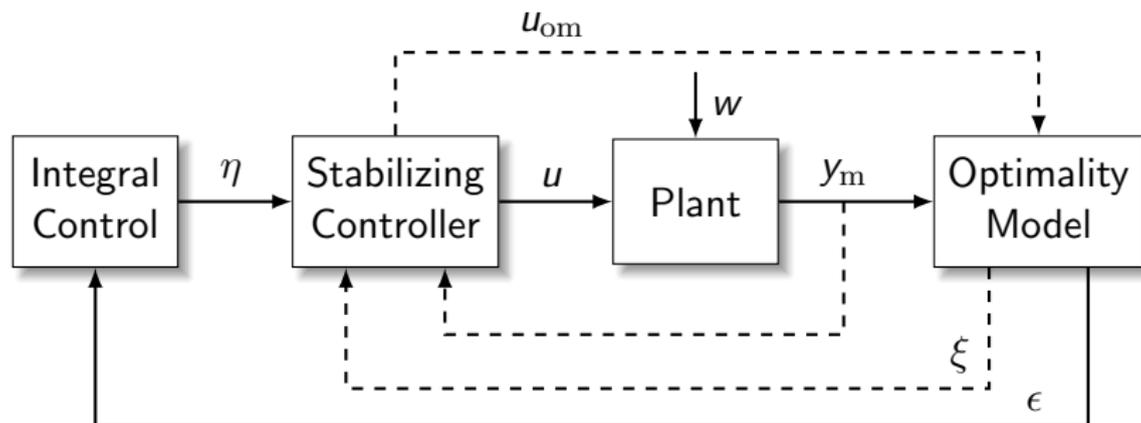
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Robustness Issues in Constructing Optimality Models

Can we implement an optimality model that is *robust* against δ ?

$$\begin{aligned} & \underset{y \in \mathbb{R}^p}{\text{minimize}} && f(y, w) \\ & \text{subject to} && y \in \bar{Y}(w, \delta) = y(w, \delta) + V(\delta) \\ & && Hy = Lw \\ & && Jy \leq Mw \end{aligned}$$

Optimality condition:

$$\nabla f(y^*, w) + J^T \nu^* \perp (V(\delta) \cap \text{null}(H))$$

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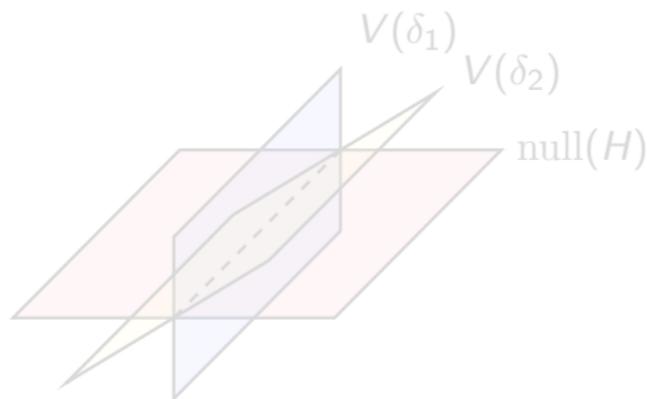
Robustness Issues (cont.)

When can an optimality model encode the gradient KKT condition?

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Robust Feasible Subspace Property

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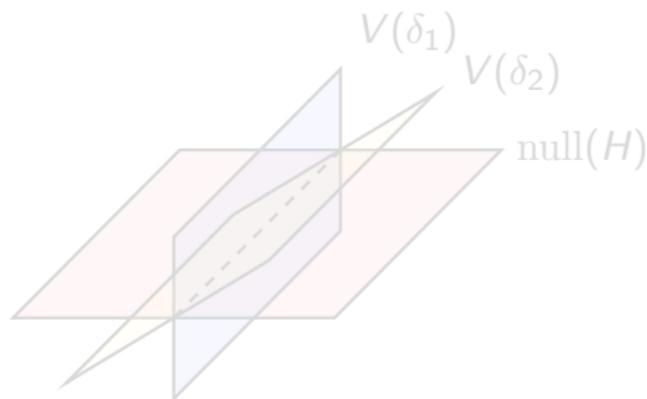
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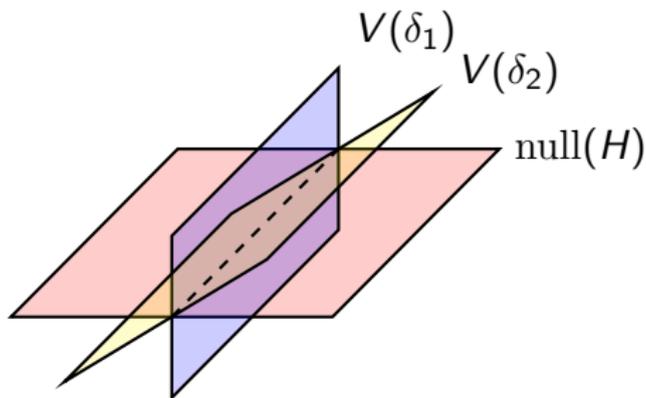
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Robust Feasible Subspace Optimality Model

If Robust Feasible Subspace property holds, then

$$\dot{\nu} = \mathbf{max}(\nu + Jy - Mw, 0) - \nu$$

$$\epsilon = \begin{bmatrix} Hy - Lw \\ \mathbf{T}_0^T(\nabla f(y, w) + J^T \nu) \end{bmatrix}$$

$$\begin{aligned} &\text{range}(\mathbf{T}_0) \\ &= V(\delta) \cap \text{null}(H) \end{aligned}$$

(Design freedom!)

is an optimality model for the LTI-Convex OSS Control Problem.

Comments:

- 1 $\mathbf{T}_0^T z$ extracts component of z in subspace $V(\delta) \cap \text{null}(H)$:

$$\epsilon_2 = 0 \quad \iff \quad \nabla f(y, w) + J^T \nu \perp V(\delta) \cap \text{null}(H)$$

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Robust Output Subspace Optimality Model

If furthermore $V(\delta)$ itself is independent of δ , then

$$\dot{\mu} = Hy - Lw$$

$$\dot{\nu} = \mathbf{max}(\nu + Jy - Mw, \mathbb{0}) - \nu$$

$$\epsilon = \mathbf{R}_0^T (\nabla g(y, w) + H^T \mu + J^T \nu)$$

$$\text{range } R_0 = V(\delta)$$

(Design freedom!)

is also an optimality model for the LTI-Convex OSS Control Problem.

- ① Can take $R_0 = I$ if $V(\delta) = \mathbb{R}^p$, which holds if

$$\begin{bmatrix} A & B \\ C & D \end{bmatrix} \text{ has full row rank} \iff \text{No transmission zeros at } s = 0$$

- ② Again, different equivalent formulations of optimization problem give different optimality models

Robust Output Subspace Optimality Model

If furthermore $V(\delta)$ itself is independent of δ , then

$$\dot{\mu} = Hy - Lw$$

$$\dot{\nu} = \mathbf{max}(\nu + Jy - Mw, \mathbb{0}) - \nu$$

$$\epsilon = \mathbf{R}_0^T (\nabla g(y, w) + H^T \mu + J^T \nu)$$

$$\text{range } R_0 = V(\delta)$$

(Design freedom!)

is also an optimality model for the LTI-Convex OSS Control Problem.

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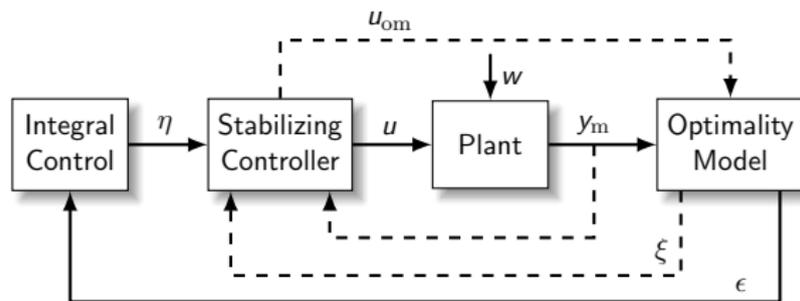
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Towards an internal model principle ...

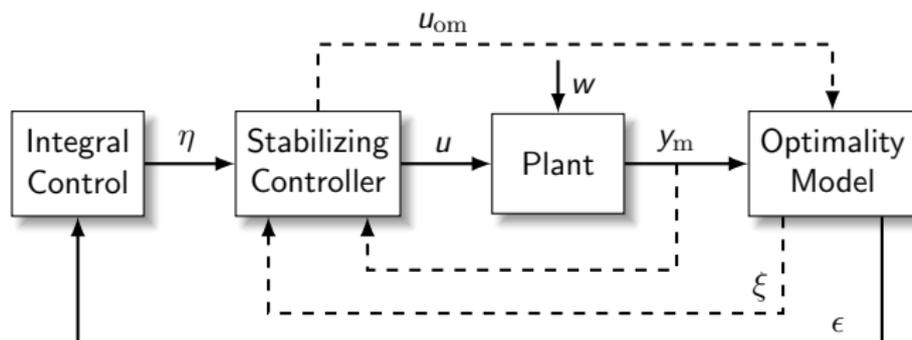
$$\epsilon = \begin{bmatrix} Hy - Lw \\ T_0^T \nabla f(y, w) \end{bmatrix}$$

$$\text{range}(T_0) = \mathcal{V}(\delta) \cap \text{null}(H)$$



Interpretation: Exact robust asymptotic optimization achieved if loop *incorporates a model of the optimal set of the optimization problem*

Stabilizer Design



Stabilizer design options:

- 1 full-order dynamic robust controller synthesis
- 2 low-gain integral control $u = -K\eta$ (Davison '76)
- 3 any heuristic, local linearized LQ design, ...

Closed-loop analysis:

- 1 Time-scale separation, robust control

Optimal Frequency Regulation Problem

- 1 Linearized dynamics of **network of generators**

$$\Delta \dot{\theta}_i = \Delta \omega_i,$$

$$M_i \Delta \dot{\omega}_i = - \sum_{j=1}^n T_{ij} (\Delta \theta_i - \Delta \theta_j) - D_i \Delta \omega_i + \Delta P_{m,i} + \Delta P_{u,i}$$

$$T_{ch,i} \Delta \dot{P}_{m,i} = -\Delta P_{m,i} + \Delta P_{g,i}$$

$$T_{g,i} \Delta \dot{P}_{g,i} = -\Delta P_{g,i} - R_{d,i}^{-1} \Delta \omega_i + \Delta P_i^{\text{ref}}.$$

- 2 Economically select equilibrium reserve powers ΔP_i^{ref} subject to balance of supply and demand (control and disturbances)

$$\text{minimize}_{\Delta P_i^{\text{ref}} \in \mathbb{R}^n} \sum_{i=1}^n J_i(P_i^{\text{ref}} + \Delta P_i^{\text{ref}})$$

$$\text{subject to } \sum_{i=1}^n (\Delta P_{u,i} + \Delta P_i^{\text{ref}}) = 0$$

Optimal Frequency Regulation Problem

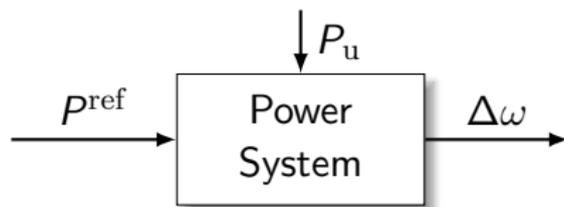
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Optimal Frequency Regulation Problem



Equivalent formulations of steady-state dispatch problem

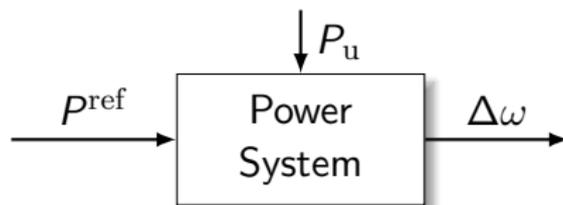
$$\begin{aligned} & \underset{P^{\text{ref}} \in \mathbb{R}^n}{\text{minimize}} && \sum_{i=1}^n J_i(P_i^{\text{ref}}) \\ & \text{subject to} && \sum_{i=1}^n (P_{u,i} + P_i^{\text{ref}}) = 0 \end{aligned}$$

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OSS Framework Recovers Recent Controllers

1 Distributed-Averaging Integral Control

$$k_i \dot{\eta}_i = -\Delta\omega_i - \sum_{j=1}^n a_{ij}(\eta_i - \eta_j), \quad P_i^{\text{ref}} = (\nabla J_i)^{-1}(\eta_i)$$

2 AGC / Gather-and-Broadcast Control

$$\dot{\eta} = \text{average}(\omega_j), \quad P_i^{\text{ref}} = (\nabla J_i)^{-1}(\eta)$$

3 Primal-dual algorithm

$$\begin{aligned} \dot{\mu}_i &= -\nabla J_i(\mu_i) - \nu, & P_i^{\text{ref}} &= \mu_i \\ \dot{\nu} &= \sum_{i=1}^n (P_{u,i} + \mu_i) \end{aligned}$$

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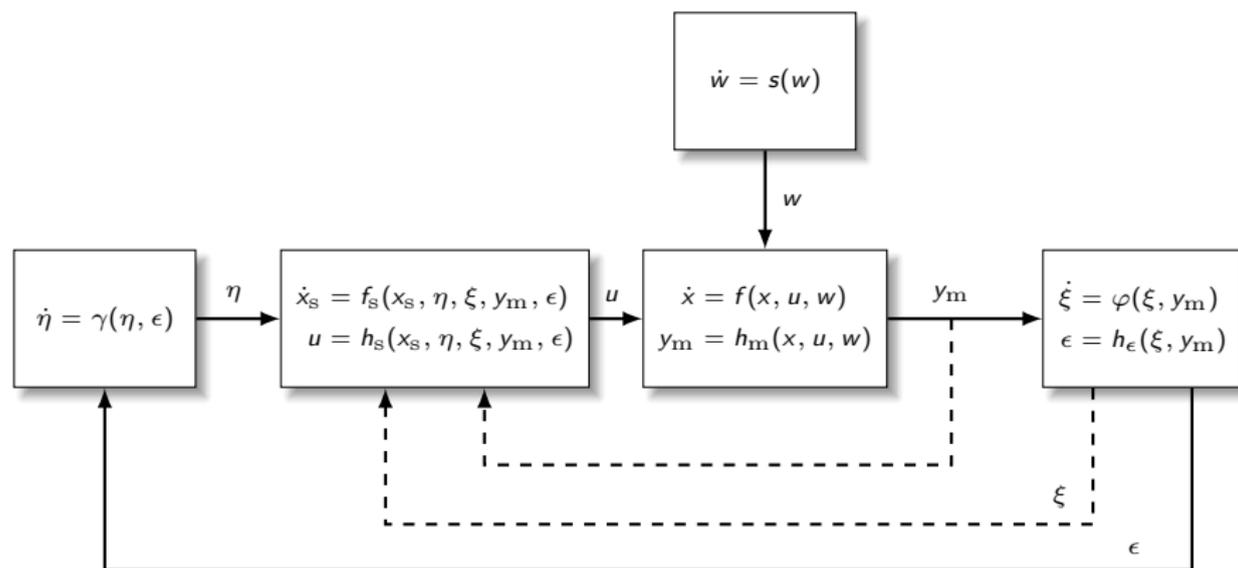
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Nonlinear OSS Control Problem

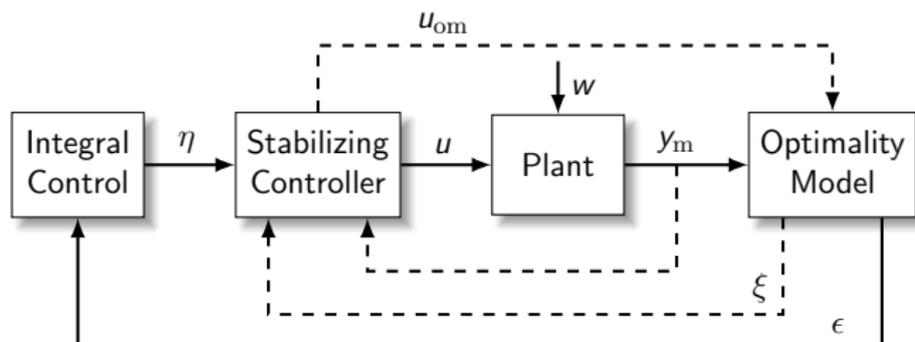
Nonlinear systems with **time-varying** disturbances in continuous or discrete-time



Conclusions

New control framework: Optimal Steady-State (OSS) Control

- 1 Robust feedback optimization of dynamic systems
- 2 Optimality model reduces OSS problem to output reg. problem



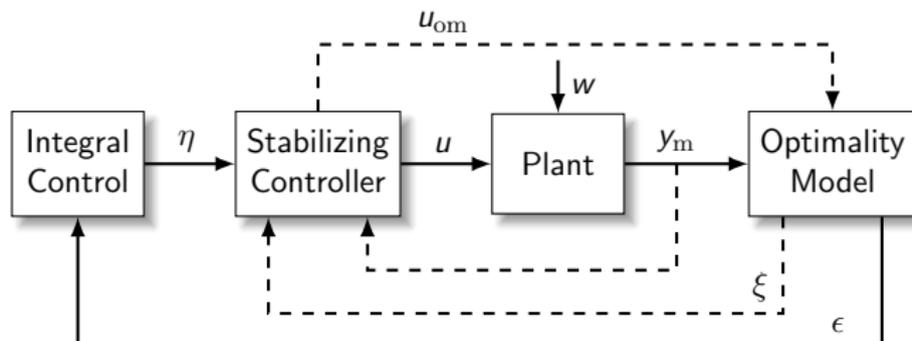
Many pieces of theory wide open ...

- 1 Sampled-data, decentralized, hierarchical, competitive, ...
- 2 Performance improvement (e.g., feedforward, anti-windup)

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The Optimal Steady-State Control Problem

Liam S. P. Lawrence *Student Member, IEEE*, John W. Simpson-Porco, *Member, IEEE*, and Enrique Mallada *Member, IEEE*

<https://arxiv.org/abs/1810.12892>



Liam S. P. Lawrence
University of Waterloo



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John's Hopkins Univ.

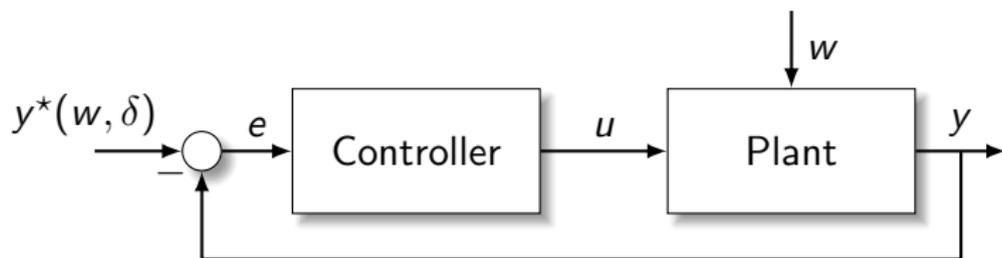
Questions



<https://ece.uwaterloo.ca/~jwsimpso/>
jwsimpson@uwaterloo.ca

appendix

Is OSS Control just a standard tracking problem?

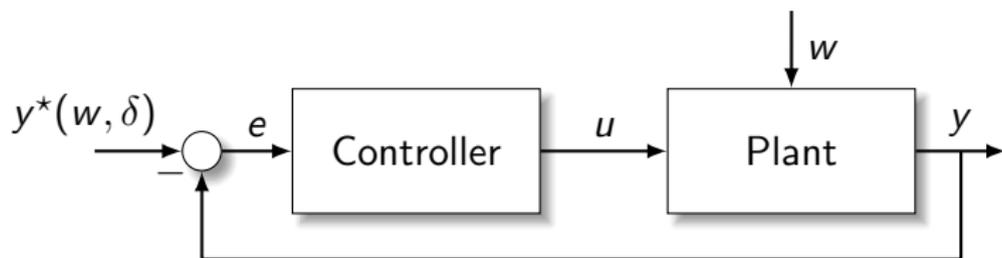


We want y to track $y^*(w, \delta)$, but two problems:

- 1 unmeasured components of w change y^*
- 2 y^* depends on uncertainty δ (relevant if $\bar{Y} \subset \mathbb{R}^P$)

Standard tracking approach **infeasible** for quickly varying $w(t)$, or large uncertainties δ , or particular choices of regulated outputs

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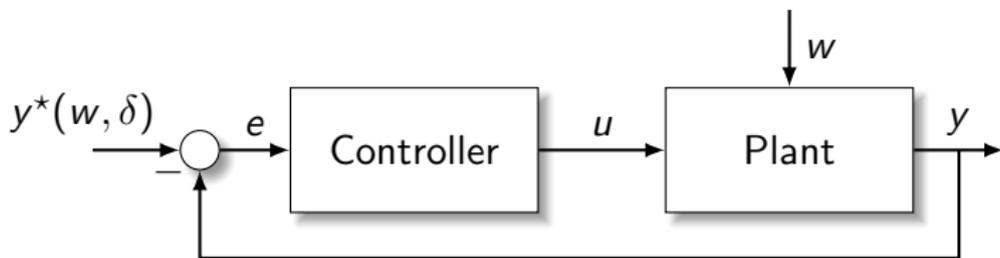


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Slide on EOA Approach . . .

Example 1: Necessity of Equilibrium Constraints

Consider the OSS control problem:

① Dynamics:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} -1 & 0 \\ 1 & -1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 1 \\ -1 \end{bmatrix} u + \begin{bmatrix} 1 \\ 1 \end{bmatrix} w$$
$$y = \begin{bmatrix} x_1 \\ u \end{bmatrix}$$

② Optimization problem:

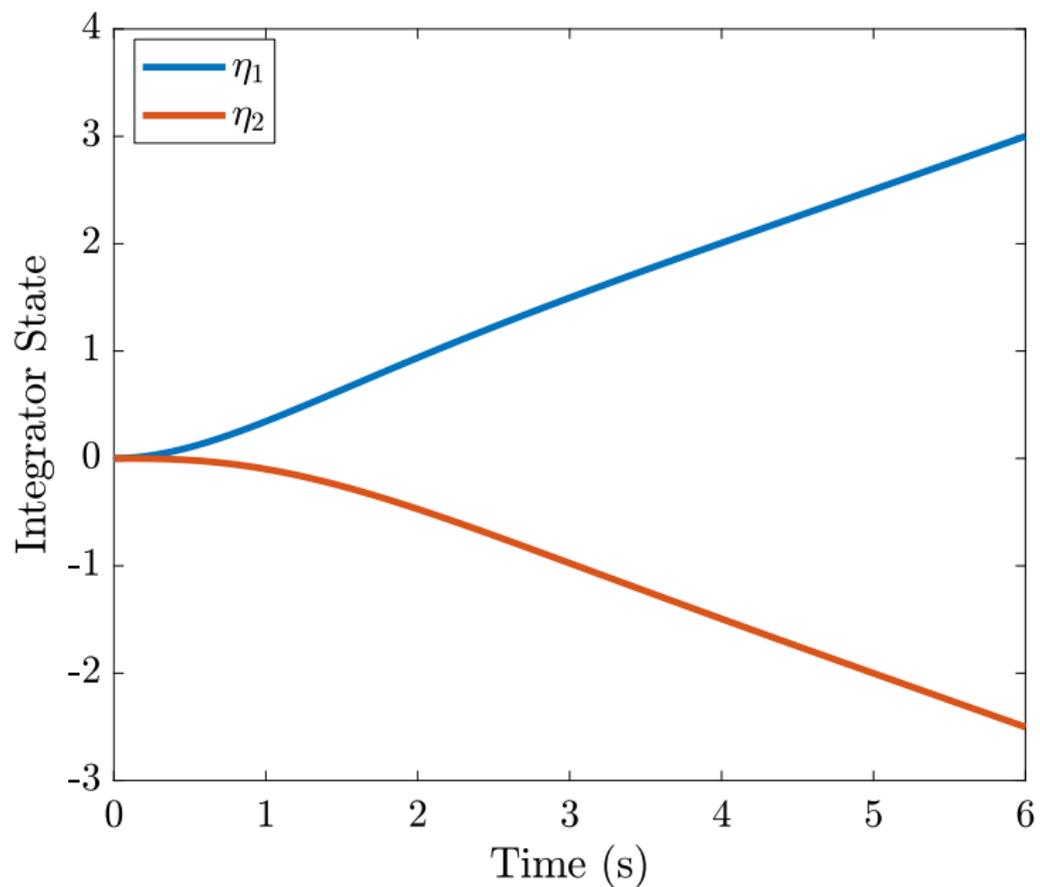
$$\underset{y \in \mathbb{R}^2}{\text{minimize}} \quad g(y) := \frac{1}{2}y_1^2 + \frac{1}{2}y_2^2$$

What happens if we omit the equilibrium constraints?

$$\dot{\eta} = \nabla g(y)$$

$$u = -K\eta$$

Example 1: Necessity of Equilibrium Constraints (cont.)



Example 2: Necessity of Robust Feasible Subspace

Consider the OSS control problem:

① Dynamics:

$$\begin{aligned} \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} &= \begin{bmatrix} -1 - \delta & 0 \\ 1 + \delta & -1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 1 \\ -1 \end{bmatrix} u + \begin{bmatrix} 1 \\ 1 \end{bmatrix} w \\ y &= \begin{bmatrix} x_1 \\ u \end{bmatrix} \end{aligned}$$

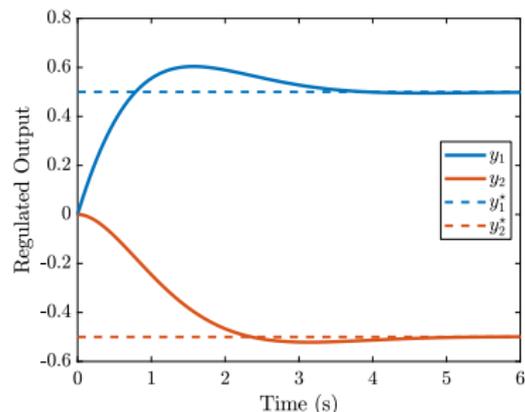
② Optimization problem:

$$\begin{aligned} &\underset{y \in \mathbb{R}^2}{\text{minimize}} && \frac{1}{2}y_1^2 + \frac{1}{2}y_2^2 \\ &\text{subject to} && y \in \bar{Y}(w, \delta) = y(w, \delta) + V(\delta) \end{aligned}$$

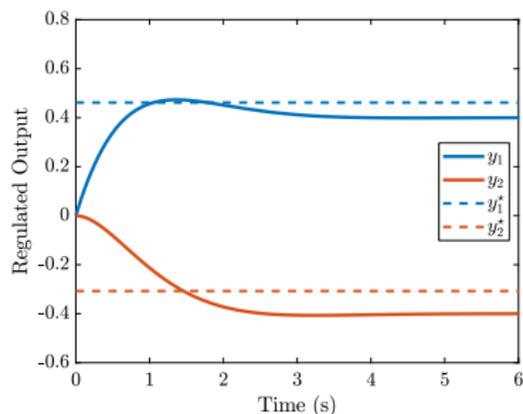
We can show $V(\delta) = \text{span} \left\{ \begin{bmatrix} 1 \\ \delta \end{bmatrix} \right\} \Rightarrow V(\delta)$ **dependent** on δ .

Example 2: Necessity of Robust Feasible Subspace (cont.)

- We apply our scheme anyway supposing $\delta = 0$
- Optimality model + integral control yields...



If $\delta = 0$ in the true plant
 \Rightarrow achieve optimal cost of 0.1538.



If $\delta = 0.5$ in the true plant
 \Rightarrow achieve sub-optimal cost of 0.1599.

OSS Control in the Literature

The OSS controller architecture found throughout the literature on **real-time optimization**.

Problem [Nelson and Mallada '18]

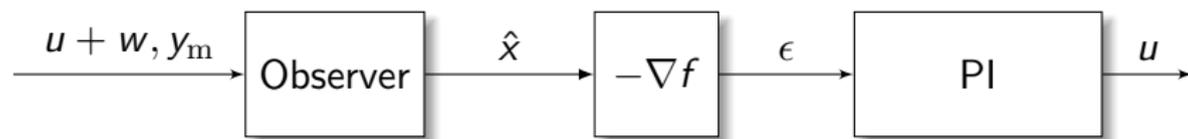
Design a feedback controller to drive the system

$$\begin{aligned}\dot{x}(t) &= Ax(t) + B(u(t) + w) \\ y_m(t) &= Cx(t) + D(u(t) + w)\end{aligned}$$

to the solution of the optimization problem

$$\underset{x \in \mathbb{R}^n}{\text{minimize}} f(x).$$

OSS Control in the Literature (cont.)



Controller Design

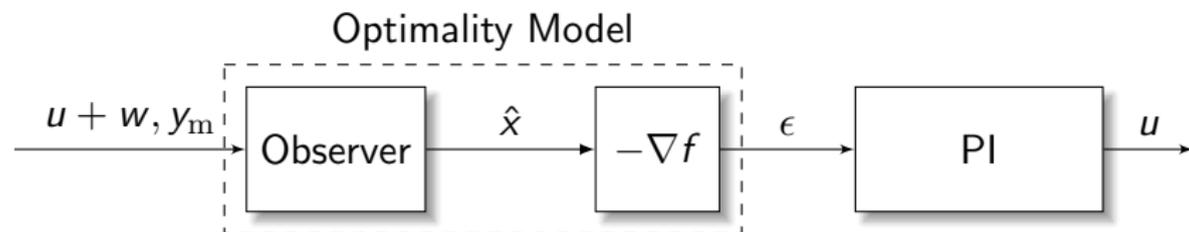
The **optimality model** is an observer with gradient output

$$\begin{aligned}\dot{\hat{x}} &= (A - LC)\hat{x} + (B - LD)(u + w) + Ly_m \\ \epsilon &= -\nabla f(\hat{x}).\end{aligned}$$

A PI controller serves as **internal model and stabilizer**

$$\dot{e}_I = \epsilon, \quad u = K_I e_I + K_P \epsilon.$$

OSS Control in the Literature (cont.)



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