

The Return of the Tuning Regulator

John W. Simpson-Porco

<https://www.control.utoronto.ca/~jwsimpson/>



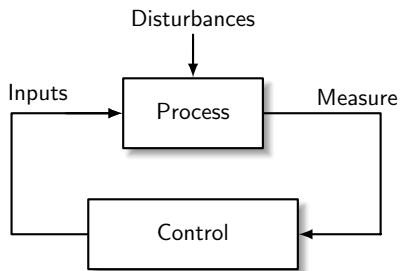
The Edward S. Rogers Sr. Department
of Electrical & Computer Engineering
UNIVERSITY OF TORONTO

9th Meeting on System and Control Theory

University of Waterloo

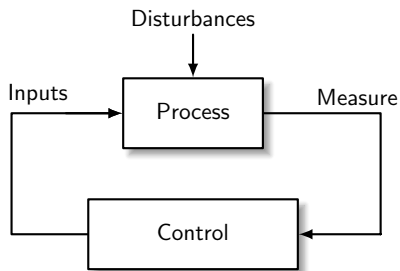
May 4, 2023

Introduction



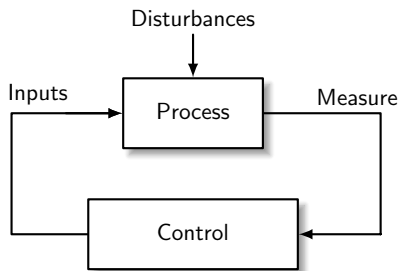
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- Design criteria
 - (i) want to track and reject
 - (ii) use minimal model info.
 - (iii) robustness \gg performance

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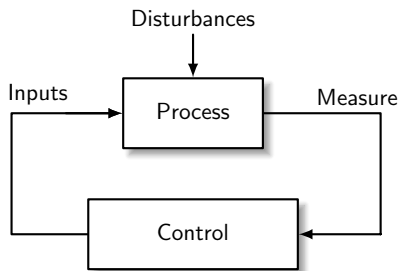
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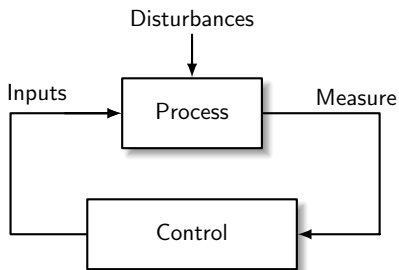
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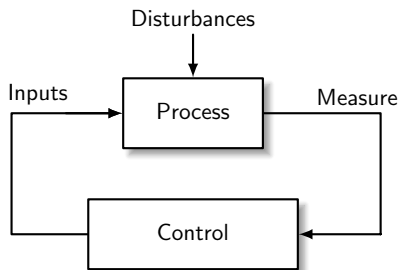
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Problem falls within the well-studied topic of **output regulation** or **servomechanism design**

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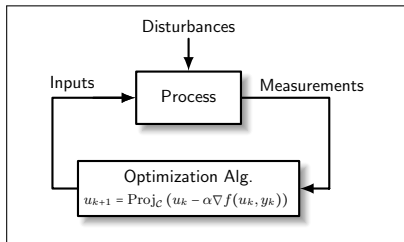
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Objective: Design low-order, easily tuned, and (nearly) model-free regulation loops for stable MIMO systems

My Motivating Applications

Feedback-Based Optimization



Analysis of Optimization Algorithms

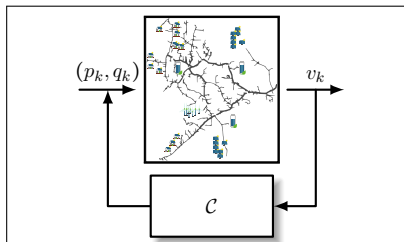
$$\underset{x \in \mathbb{R}^n}{\text{minimize}} \quad f(x) \quad \text{s.t.} \quad Ax = b$$

$$x_{k+1} = x_k - \alpha (\nabla f(x_k) + A^T \lambda_k)$$

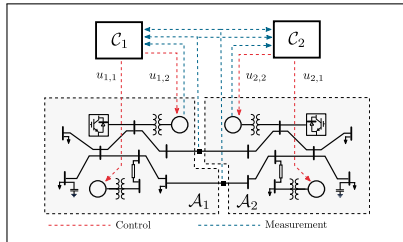
$$\tilde{x}_k = x_k + \gamma (x_{k+1} - x_k)$$

$$\lambda_{k+1} = \lambda_k + \beta (A \tilde{x}_k - b),$$

Renewable Energy Integration



Next-Generation Grid Control



This Talk

4148

IEEE TRANSACTIONS ON AUTOMATIC CONTROL, VOL. 66, NO. 9, SEPTEMBER 2021



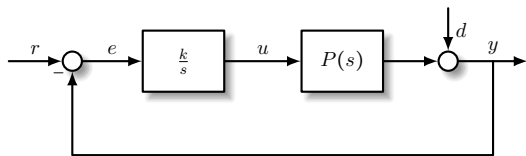
Analysis and Synthesis of Low-Gain Integral Controllers for Nonlinear Systems

John W. Simpson-Porco , *Member, IEEE*

Data-Driven Output Regulation using Single-Gain Tuning Regulators

Liangjie Chen and John W. Simpson-Porco

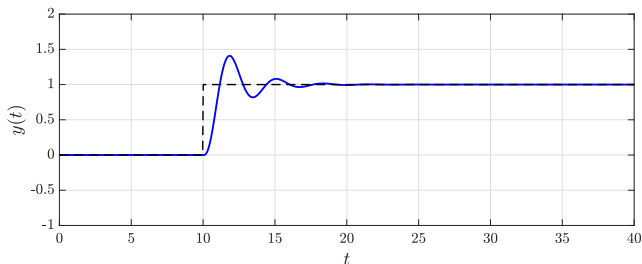
Fundamentals of Integral Control



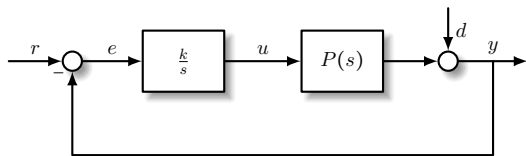
Assume P stable, $P(0) \neq 0$. What are the basic facts and tuning principles?

The Integral Control Dichotomy: Either

(a) the closed-loop is BIBO stable **and** $\lim_{t \rightarrow \infty} e(t) = 0$, or



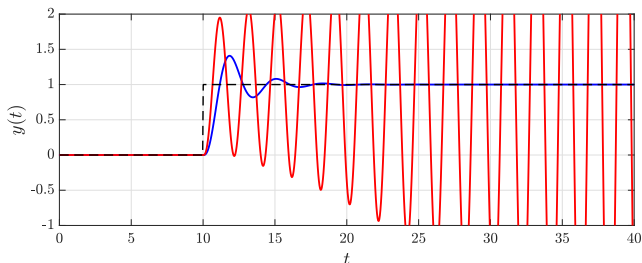
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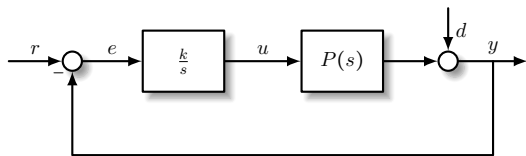
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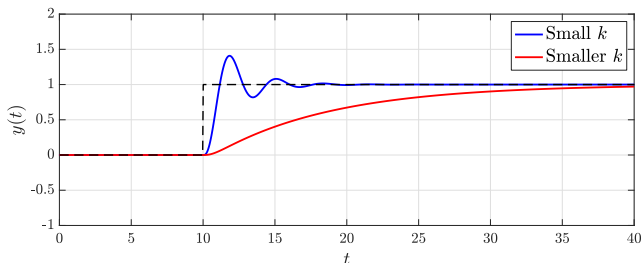
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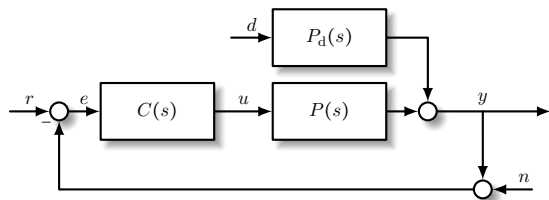
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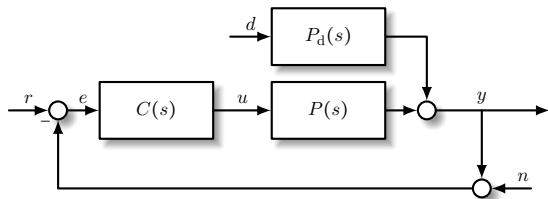


Low-Gain Integral Control of LTI Systems



- $P, P_d =$ LTI
exp. stable
- $d, r =$ constant

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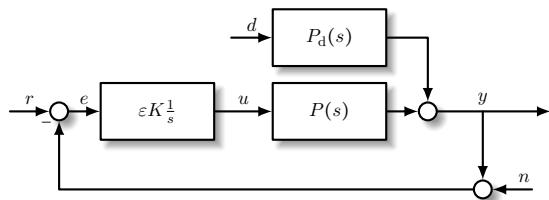
Multivariable Tuning Regulators:
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EDWARD J. DAVISON, MEMBER, IEEE

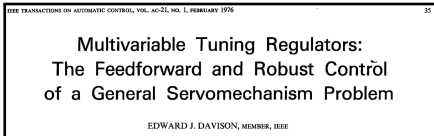


$$\dot{\eta} = e$$
$$u = -\varepsilon K \eta$$
$$K = P(0)^\dagger$$

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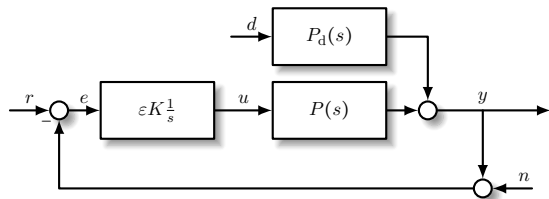


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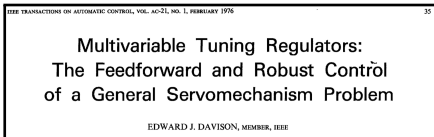


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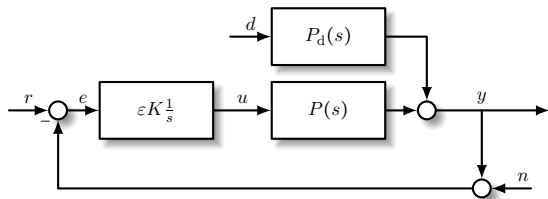
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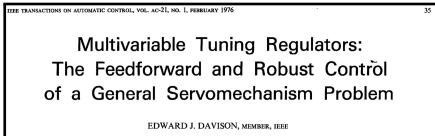
$$-P(0)K \text{ Hurwitz} \implies \exists \varepsilon^* > 0 \text{ s.t. } \forall \varepsilon \in (0, \varepsilon^*)$$

$$\text{C.L.S. exp. stable \& } e(t) \rightarrow 0$$

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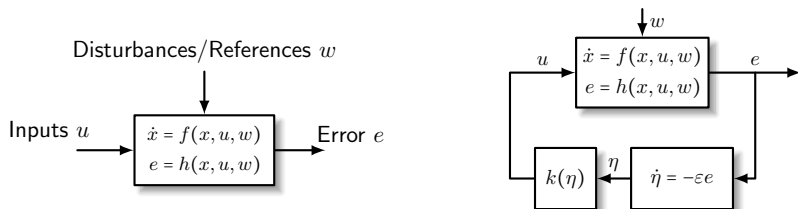


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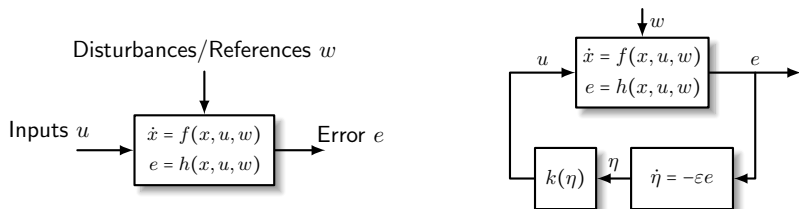
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Only required model information is the DC gain!

Low-Gain Integral Control of Nonlinear Systems



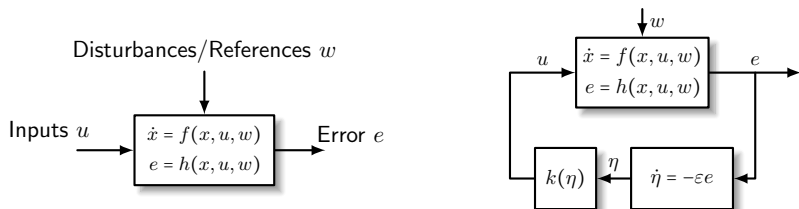
Low-Gain Integral Control of Nonlinear Systems



Assumptions: There exist state and input sets \mathcal{X} and \mathcal{I} such that

- 1 **Model Regularity:** f and $\frac{\partial f}{\partial x}$ are Lipschitz on \mathcal{X} uniformly in $(u, w) \in \mathcal{I}$,
- 2 **Steady-State:** $\exists \pi_x : \mathcal{I} \rightarrow \mathcal{X}$ s.t. $0 = f(\pi_x(u, w), u, w)$ for all $(u, w) \in \mathcal{I}$,
- 3 **Stability:** $\bar{x} = \pi_x(u, w)$ is locally exponentially stable, uniformly in $(u, w) \in \mathcal{I}$,
- 4 **Integral Gain:** k is class C^1 and Lipschitz.

Low-Gain Integral Control of Nonlinear Systems



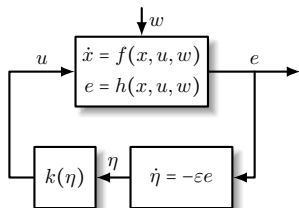
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Equilibrium I/O Map:

$$\begin{aligned} \pi(\bar{u}, w) &\triangleq h(\pi_x(\bar{u}, w), \bar{u}, w) \\ &= P(0)\bar{u} + P_w(0)w \text{ for LTI} \end{aligned}$$

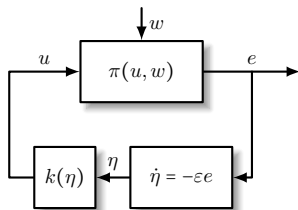
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Equivalent

- 1 ε is **small**
- 2 integral action is **slow**
- 3 process is **fast**

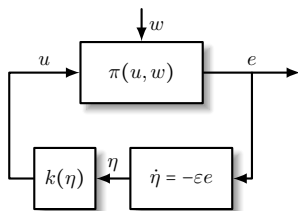
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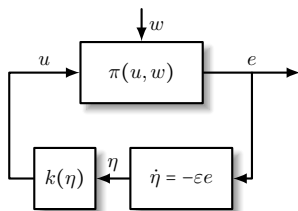
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Reduced Dynamics

$$\begin{aligned}\frac{1}{\varepsilon} \dot{\eta} &= -\pi(k(\eta), w) \triangleq F_w(\eta) \\ &= -P(0)K\eta - P_w(0)w\end{aligned}$$

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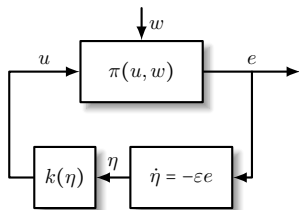
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Reduced Dynamics

Infinitesimally Contracting \implies

$$\sup_{\eta, w} \mu \left(\frac{\partial F_w}{\partial \eta}(\eta) \right) < 0$$

$$\exists \varepsilon^* > 0 \text{ s.t. } \forall \varepsilon \in (0, \varepsilon^*)$$

C.L.S. has exp. stable equil.

$$(\bar{x}, \bar{\eta}) \ \& \ e(t) \rightarrow 0$$

A Brief Aside on Contraction

Contraction is an incremental stability concept which furnishes nonlinear systems $\dot{x} = f(x)$ with linear-like stability properties.

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- Given norm $\|\cdot\|$ on \mathbb{R}^n , the **matrix measure** or **induced log norm** is the map $\mu : \mathbb{R}^{n \times n} \rightarrow \mathbb{R}$ defined by $\mu(A) = \lim_{h \rightarrow 0^+} \frac{\|I+hA\| - 1}{h}$

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- **Example:** If $\|x\| = \sqrt{x^\top P x}$, then $\mu(A) = \lambda_{\max}(A^\top P + P A)$

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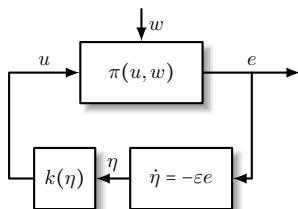
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Suppose $f \in C^1$ is globally **infinitesimally contracting**, i.e.,

$$\sup_{x \in \mathbb{R}^n} \mu\left(\frac{\partial f}{\partial x}(x)\right) < 0.$$

Then $\dot{x} = f(x)$ has a **unique globally exponentially stable** equilibrium point \bar{x} .

Low-Gain Integral Control of Nonlinear Systems



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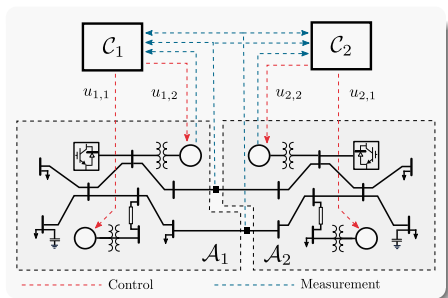
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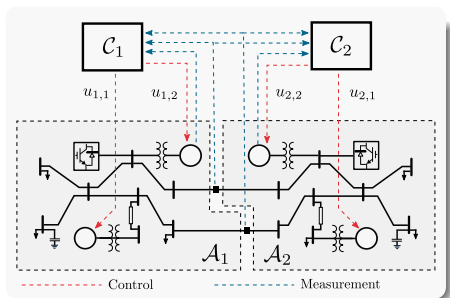
Application: Automatic Generation Control

Rebalancing supply and demand in interconnected systems



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Rebalancing supply and demand in interconnected systems



- Area Control Error (ACE)

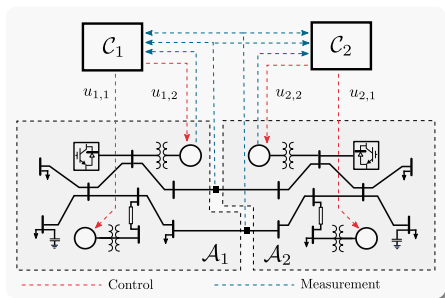
$$ACE_k(t) := \underbrace{\Delta N I_k(t)}_{\text{Net Interchange}} + \underbrace{b_k \Delta f_k(t)}_{\text{Frequency Biasing}}$$

- Area-based integral controller

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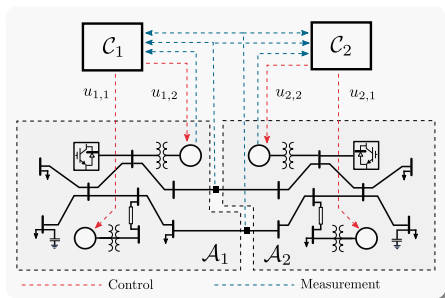
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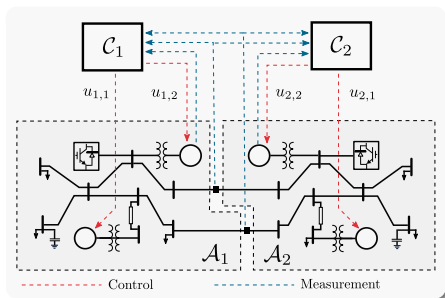
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- Definitive analysis enabled by previous result, and some advances in theory of diagonal stability; see paper in Trans. Control Network Systems

Designing Low-Gain Integral Controllers

Contraction of $\dot{\eta} = -\pi(k(\eta), w)$ is the design goal.

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If π_1 is surjective, choose $k = \pi_1^\dagger$ (cf. $K = P(0)^\dagger$)

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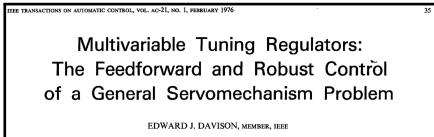
If π_1 is surjective, choose $k = \pi_1^\dagger$ (cf. $K = P(0)^\dagger$)

- (ii) **Robust Design:** If you can represent/encapsulate $\pi(u, w)$ in a linear fractional model

$$\begin{aligned} e &= Fu + Gp + E_1w \\ \mathbf{q} &= Hu + Jp + E_2w \end{aligned} \quad \mathbf{p} = \Delta(\mathbf{q})$$

where Δ satisfies **incremental quadratic constraints**; can immediately apply SDP-based robust/optimal design methods

Davison's Tuning Regulator



Tuning Regulators generalize low-gain integral control to more general **multi-modal disturbances**.

Davison's Tuning Regulator

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Multivariable Tuning Regulators:
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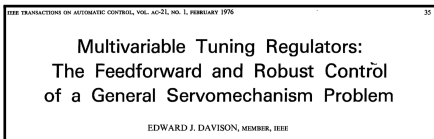


Tuning Regulators generalize low-gain integral control to more general **multi-modal disturbances**.

$$\mathcal{P} : \begin{aligned} \dot{x} &= Ax + Bu + B_d d \\ e &= Cx + Du + D_d d \\ \text{eig}(A) &\subset \mathbb{C}_- \\ \text{T.F. } \hat{P}(s) &\in \mathbb{C}^{r \times m} \end{aligned}$$

$$\mathcal{E} : \begin{aligned} \dot{w} &= Sw \\ d &= Ew \\ \text{eig}(S) &\subset j\mathbb{R} \end{aligned}$$

Davison's Tuning Regulator



Tuning Regulators generalize low-gain integral control to more general **multi-modal disturbances**.

$$\mathcal{P} : \begin{aligned} \dot{x} &= Ax + Bu + B_d d \\ e &= Cx + Du + D_d d \\ \text{eig}(A) &\subset \mathbb{C}_- \\ \text{T.F. } \hat{P}(s) &\in \mathbb{C}^{r \times m} \end{aligned}$$

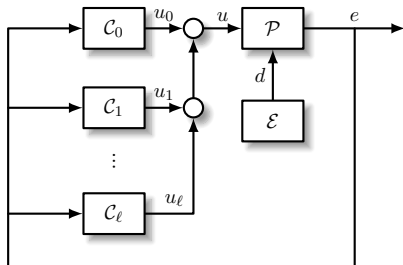
$$\mathcal{E} : \begin{aligned} \dot{w} &= Sw \\ d &= Ew \\ \text{eig}(S) &\subset j\mathbb{R} \end{aligned}$$

For Simplicity: $\text{MinPoly}_S(s) = s(s^2 + \omega_1^2) \cdots (s^2 + \omega_\ell^2)$

Necessarily: $\text{rank } \hat{P}(\lambda) = r$ for all $\lambda \in \text{eig}(S)$.

Architecture of Davison's Tuning Regulator

Idea: Sub-controllers handle each disturbance mode

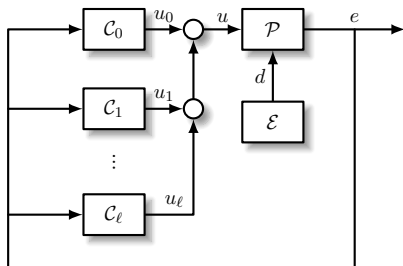


$$\mathcal{C}_k : \begin{aligned} \dot{\eta}_k &= \Phi_k \eta_k + G_k e \\ u_k &= -\epsilon_k F_k \eta_k \end{aligned}$$

$$\Phi_0 = \mathbf{0}_r, \quad \Phi_k = \begin{bmatrix} 0 & 0 \\ -\omega_k^2 & 0 \end{bmatrix} \otimes I_r$$
$$G_0 = I_r, \quad G_k = \begin{bmatrix} 0 \\ 1 \end{bmatrix} \otimes I_r$$

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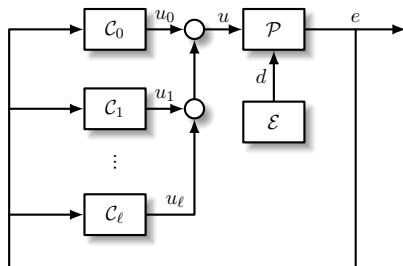
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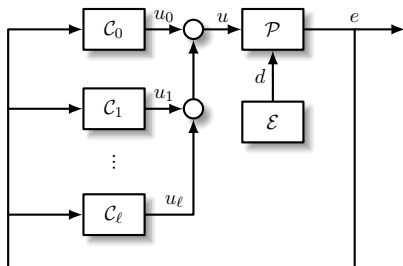
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- ☹☹: Must **re-identify** freq. response after each tuning step

The Single-Gain Tuning Regulator

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- **Definition:** $\mathcal{A}(\epsilon)$ is *low-gain stable* if there exist $c, \epsilon^* > 0$ such that $\text{Re}(\lambda) \leq -c\epsilon$ for all $\lambda \in \text{eig}(\mathcal{A}(\epsilon))$ and all $\epsilon \in [0, \epsilon^*]$.

Reduction of Stability Analysis Problem

$$\mathcal{A}(\epsilon) := \begin{bmatrix} A & -BF(\epsilon) \\ GC & \Phi - GDF(\epsilon) \end{bmatrix} \iff \mathcal{A}_{\text{red}}(\epsilon) = \Phi - G\mathcal{L}(F(\epsilon))$$

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where $\mathcal{L} : \mathbb{R}^{m \times r q} \rightarrow \mathbb{R}^{r \times r q}$

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Idea: Place poles of (Φ, G) with feedback gain $K(\epsilon)$, then solve linear operator equation $\mathcal{L}(F) = K$ for F .

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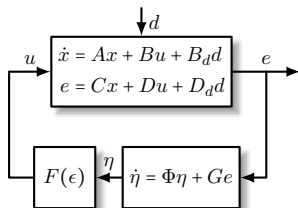
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$$\begin{aligned} \mathcal{L}(F) &= C \text{Syl}^{-1}(BF) + DF \\ \text{Syl}(X) &= X\Phi - AX \end{aligned}$$

$$\mathcal{L}(F) = \hat{P}(0)F \mathbf{X}_0 + 2 \sum_{k=1}^{\ell} \text{Re}\{\hat{P}(\mathbf{j}\omega_k)F \mathbf{X}_k\}.$$

where matrix \mathbf{X}_k comes from eigen. decomp. of Φ .

Summary

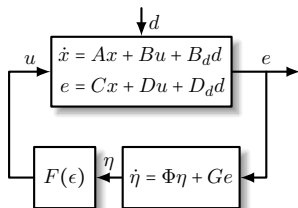


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Design Procedure:

- 1 Design $K(\epsilon)$ such that $\Phi - GK(\epsilon)$ is low-gain stable; this is a **state-feedback problem** (pole placement, robust, optimal, ...)
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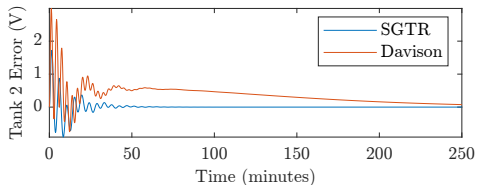
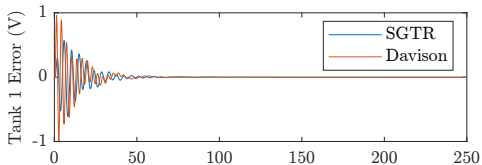
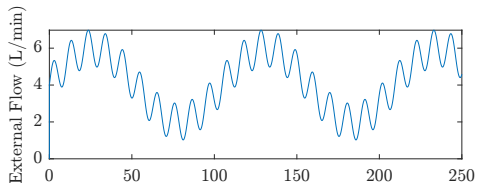
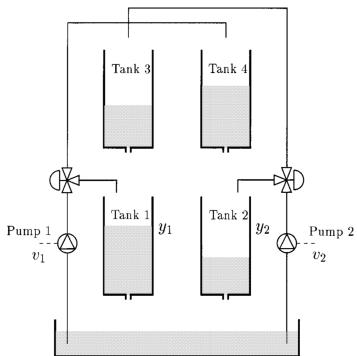
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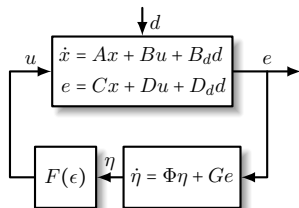
Full design authority over slow time-scale dynamics.

Example: Four-Tank Process (Johannson, TCST, 2000)



Conclusions

- 1 Low-complexity easily-tuned robust data-driven design
- 2 Integral controllers for nonlinear systems
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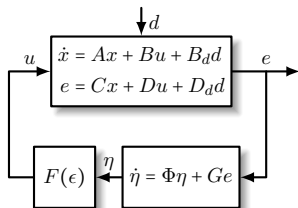


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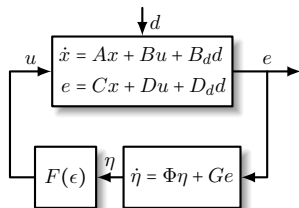


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Note: Open PhD position at University of Toronto for Fall 2023, focusing on data-driven control and estimation for energy systems.

Questions



The Edward S. Rogers Sr. Department
of Electrical & Computer Engineering
UNIVERSITY OF TORONTO

<https://www.control.utoronto.ca/~jwsimpson/>
jwsimpson@ece.utoronto.ca